Lecture -3 Fourier Transform Properties

Fourier Transform Properties and Examples

Objectives:

- 1. Properties of a Fourier transform
 - Linearity & time shifts
 - Differentiation
 - **Convolution** in the frequency domain

- While the Fourier series/transform is very important for representing a signal in the frequency domain, it is also important for calculating a system's response (convolution).
- A system's transfer function is the Fourier transform of its impulse response
- Fourier transform of a signal's derivative is multiplication in the frequency domain: jωX(jω)
- Convolution in the time domain is given by multiplication in the frequency domain (similar idea to log transformations)

Review: Fourier Transform

A CT signal x(t) and its frequency domain, Fourier transform signal, $X(j\omega)$, are related by

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt \qquad \text{analysis}$$

synthesis

 $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$ This is denoted by:

$$x(t) \stackrel{r}{\leftrightarrow} X(j\omega)$$

For example:

$$e^{-at}u(t) \stackrel{F}{\leftrightarrow} \frac{1}{a+j\omega}$$

Often you have tables for common Fourier transforms

The Fourier transform, $X(j\omega)$, represents the **frequency content** of x(t).

It exists either when x(t)->0 as |t|->∞ or when x(t) is periodic (it generalizes the Fourier series)

Linearity of the Fourier Transform

The Fourier transform is a **linear function** of *x*(*t*)

$$x_{1}(t) \stackrel{F}{\leftrightarrow} X_{1}(j\omega)$$

$$x_{2}(t) \stackrel{F}{\leftrightarrow} X_{2}(j\omega)$$

$$ax_{1}(t) + bx_{2}(t) \stackrel{F}{\leftrightarrow} aX_{1}(j\omega) + bX_{2}(j\omega)$$

- This follows directly from the definition of the Fourier transform (as the integral operator is linear) & it easily extends to an arbitrary number of signals
- Like impulses/convolution, if we know the Fourier transform of simple signals, we can calculate the Fourier transform of more complex signals which are a linear combination of the simple signals

Fourier Transform of a Time Shifted Signal

We'll show that a Fourier transform of a signal which has a **simple time shift** is:

 $F\{x(t-t_0)\} = e^{-j\omega t_0} X(j\omega)$

i.e. the original Fourier transform but shifted in phase by $-\omega t_0$

Proof

Consider the Fourier transform synthesis equation:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$
$$x(t-t_0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega(t-t_0)} d\omega$$
$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \left(e^{-j\omega t_0} X(j\omega) \right) e^{j\omega t} d\omega$$

but this is the synthesis equation for the Fourier transform $e^{-j\omega_0 t}X(j\omega)$

Example: Linearity & Time Shift

Consider the signal (linear sum of two time shifted rectangular pulses)

 $x(t) = 0.5x_1(t - 2.5) + x_2(t - 2.5)$

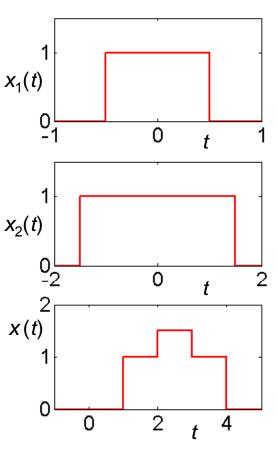
where $x_1(t)$ is of width 1, $x_2(t)$ is of width 3, centred on zero (see figures)

Using the FT of a rectangular pulse L10S7

 $X_{1}(j\omega) = \frac{2\sin(\omega/2)}{\omega}$ $X_{2}(j\omega) = \frac{2\sin(3\omega/2)}{\omega}$

Then using the **linearity** and **time shift** Fourier transform properties

$$X(j\omega) = e^{-j5\omega/2} \left(\frac{\sin(\omega/2) + 2\sin(3\omega/2)}{\omega} \right)$$



Fourier Transform of a Derivative

By differentiating both sides of the Fourier transform synthesis equation with respect to *t*.

$$\frac{dx(t)}{dt} = \frac{1}{2\pi} \int_{-\infty}^{\infty} j\omega X(j\omega) e^{j\omega t} d\omega$$

Therefore noting that this is the synthesis equation for the Fourier transform $j\omega X(j\omega)$

$$\frac{dx(t)}{dt} \stackrel{F}{\longleftrightarrow} j\omega X(j\omega)$$

This is very important, because it replaces **differentiation** in the **time domain** with **multiplication** (by $j\omega$) in the **frequency domain**.

We can **solve ODEs** in the **frequency domain** using **algebraic** operations (see next slides)

Convolution in the Frequency Domain

We can easily solve ODEs in the frequency domain: $y(t) = h(t) * x(t) \leftrightarrow Y(j\omega) = H(j\omega)X(j\omega)$

Therefore, to apply **convolution in the frequency domain**, we just have to **multiply** the **two Fourier Transforms**.

- To solve for the differential/convolution equation using Fourier transforms:
- 1. Calculate **Fourier transforms** of x(t) and h(t): $X(j\omega)$ by $H(j\omega)$
- **2.** Multiply $H(j\omega)$ by $X(j\omega)$ to obtain $Y(j\omega)$
- 3. Calculate the **inverse Fourier transform** of $Y(j\omega)$

H(jω) is the LTI system's transfer function which is the Fourier transform of the impulse response, h(t). Very important in the remainder of the course (using Laplace transforms)
This result is proven in the appendix

Example 1: Solving a First Order ODE

Calculate the response of a CT LTI system with impulse response:

$$h(t) = e^{-bt}u(t) \qquad b > 0$$

to the input signal:

$$x(t) = e^{-at}u(t) \qquad a > 0$$

Taking Fourier transforms of both signals:

$$H(j\omega) = \frac{1}{b+j\omega}, \quad X(j\omega) = \frac{1}{a+j\omega}$$

gives the overall frequency response:

$$Y(j\omega) = \frac{1}{(b+j\omega)(a+j\omega)}$$

to convert this to the time domain, express as partial fractions:

$$Y(j\omega) = \frac{1}{b-a} \left(\frac{1}{(a+j\omega)} - \frac{1}{(b+j\omega)} \right)$$

assume b≠a

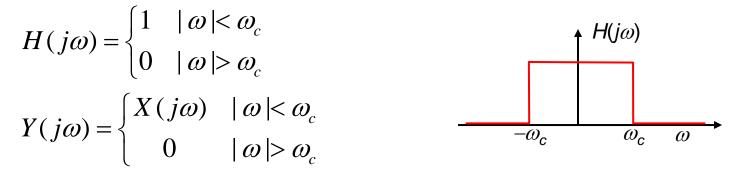
Therefore, the CT system response is:

$$y(t) = \frac{1}{b-a} \left(e^{-at} u(t) - e^{-bt} u(t) \right)$$

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Example 2: Design a Low Pass Filter

Consider an ideal **low pass filter** in frequency domain:



The filter's impulse response is the inverse Fourier transform

$$h(t) = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega t} d\omega = \frac{\sin(\omega_c t)}{\pi t}$$

which is an ideal low pass CT filter. However it is non-causal, so this cannot be manufactured exactly & the time-domain oscillations may be undesirable

We need to approximate this filter with a causal system such as 1^{st} order LTI system impulse response {*h*(*t*), *H*(*j* ω)}:

$$a^{-1}\frac{\partial y(t)}{\partial t} + y(t) = x(t), \qquad e^{-at}u(t) \stackrel{F}{\leftrightarrow} \frac{1}{a+j\omega}$$

Lecture 11: Summary

The Fourier transform is widely used for designing **filters**. You can design systems with reject high frequency noise and just retain the low frequency components. This is natural to describe in the **frequency domain**.

Important properties of the Fourier transform are:

- **1. Linearity** and time shifts $ax(t) + by(t) \leftrightarrow aX(j\omega) + bY(j\omega)$
- 2. Differentiation
- 3. Convolution

 $\frac{dx(t)}{dt} \stackrel{F}{\leftrightarrow} j\omega X(j\omega)$ $y(t) = h(t) * x(t) \stackrel{F}{\leftrightarrow} Y(j\omega) = H(j\omega)X(j\omega)$

Some operations are **simplified** in the frequency domain, but there are a number of signals for which the Fourier transform does not exist – this leads naturally onto **Laplace transforms**. Similar properties hold for Laplace transforms & the Laplace transform is widely used in engineering analysis.

Lecture 11: Exercises

Theory

1. Using linearity & time shift calculate the Fourier transform of

$$x(t) = 5e^{-3(t-1)}u(t-1) + 7e^{-3(t-2)}u(t-2)$$

- 2. Use the FT derivative relationship (S7) and the Fourier series/transform expression for $sin(\omega_0 t)$ (L10-S3) to evaluate the FT of $cos(\omega_0 t)$.
- 3. Calculate the FTs of the systems' impulse responses

a)
$$\frac{\partial y(t)}{\partial t} + 3y(t) = x(t)$$
 b) $3\frac{\partial y(t)}{\partial t} + y(t) = x(t)$

4. Calculate the system responses in Q3 when the following input signal is applied $x(t) = e^{-5t}u(t)$

Matlab/Simulink

- 5. Verify the answer to Q1 using the Fourier transform toolbox in Matlab
- 6. Verify Q3 and Q4 in Simulink
- 7. Simulate a first order system in Simulink and input a series of sinusoidal signals with different frequencies. How does the response depend on the input frequency (S12)?

Appendix: Proof of Convolution Property

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

Taking Fourier transforms gives:

$$Y(j\omega) = \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau \right) e^{-j\omega t} dt$$

Interchanging the order of integration, we have

$$Y(j\omega) = \int_{-\infty}^{\infty} x(\tau) \left(\int_{-\infty}^{\infty} h(t-\tau) e^{-j\omega t} dt \right) d\tau$$

By the time shift property, the bracketed term is $e^{-j\omega\tau}H(j\omega)$, so

$$Y(j\omega) = \int_{-\infty}^{\infty} x(\tau)e^{-j\omega\tau}H(j\omega)d\tau$$
$$= H(j\omega)\int_{-\infty}^{\infty} x(\tau)e^{-j\omega\tau}d\tau$$
$$= H(j\omega)X(j\omega)$$

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